Multiloop and Multivariable Control

• Process Interactions and Control Loop Interactions
• Pairing of Controlled and Manipulated Variables
• Singular Value Analysis
• Tuning of Multiloop PID Control Systems
• Decoupling and Multivariable Control Strategies
Control of Multivariable Processes

• Control systems that have only one controlled variable and one manipulated variable.
  ➢ Single-input, single-output (SISO) control system
  ➢ Single-loop control system

• In practical control problems there typically are a number of process variables which must be controlled and a number of variables which can be manipulated.
  ➢ Multi-input, multi-output (MIMO) control system

Example: product quality and throughput must usually be controlled.

Note the "process interactions" between controlled and manipulated variables.
Several simple physical examples

Process interactions:
Each manipulated variable can affect both controlled variables
Multiloop and Multivariable Control

SISO

(a) Single-input, single-output process with multiple disturbances

MIMO

(b) Multiple-input, multiple-output process (2x2)

(c) Multiple-input, multiple-output process (n x n)
In this chapter we will be concerned with characterizing process interactions and selecting an appropriate multiloop control configuration.

If process interactions are significant, even the best multiloop control system may not provide satisfactory control.

In these situations there are incentives for considering multivariable control strategies.

**Definitions:**

- **Multiloop control:** Each manipulated variable depends on only a single controlled variable, i.e., a set of conventional feedback controllers.

- **Multivariable Control:** Each manipulated variable can depend on two or more of the controlled variables.

**Examples:** decoupling control, model predictive control
Multiloop Control Strategy

- Typical industrial approach
- Consists of using several standard FB controllers (e.g., PID), one for each controlled variable.

- Control system design
  1. Select controlled and manipulated variables.
  2. Select pairing of controlled and manipulated variables.
  3. Specify types of FB controllers.

Example: 2 x 2 system

Two possible controller pairings:

- \( U_1 \) with \( Y_1 \), \( U_2 \) with \( Y_2 \)  (1-1/2-2 pairing)
- or
  - \( U_1 \) with \( Y_2 \), \( U_2 \) with \( Y_1 \)  (1-2/2-1 pairing)

Note: For \( n \times n \) system, \( n! \) possible pairing configurations.
Process Interactions

Transfer Function Model (2 x 2 system)

- Two controlled variables and two manipulated variables (4 transfer functions required)

\[
\frac{Y_1(s)}{U_1(s)} = G_{p11}(s), \quad \frac{Y_1(s)}{U_2(s)} = G_{p12}(s) \\
\frac{Y_2(s)}{U_1(s)} = G_{p21}(s), \quad \frac{Y_2(s)}{U_2(s)} = G_{p22}(s)
\]

(18 – 1)

- Thus, the input-output relations for the process can be written as:

\[
Y_1(s) = G_{p11}(s)U_1(s) + G_{p12}(s)U_2(s) \quad (18 – 2)
\]

\[
Y_2(s) = G_{p21}(s)U_1(s) + G_{p22}(s)U_2(s) \quad (18 – 3)
\]
In vector-matrix notation as

\[
Y(s) = G_p(s)U(s) \quad (18-4)
\]

where \( Y(s) \) and \( U(s) \) are vectors

\[
Y(s) = \begin{bmatrix} Y_1(s) \\ Y_2(s) \end{bmatrix} \quad \quad U(s) = \begin{bmatrix} U_1(s) \\ U_2(s) \end{bmatrix} \quad (18-5)
\]

And \( G_p(s) \) is the transfer function matrix for the process

\[
G_p(s) = \begin{bmatrix} G_{p11}(s) & G_{p12}(s) \\ G_{p21}(s) & G_{p22}(s) \end{bmatrix} \quad (18-6)
\]

The steady-state process transfer matrix \((s=0)\) is called the process gain matrix \( K \)

\[
K = \begin{bmatrix} G_{p11}(0) & G_{p12}(0) \\ G_{p21}(0) & G_{p22}(0) \end{bmatrix} = \begin{bmatrix} K_{11} & K_{12} \\ K_{21} & K_{22} \end{bmatrix}
\]
Block Diagram for 2x2 Multiloop Control

(a) 1-1/2-2 controller pairing

(b) 1-2/2-1 controller pairing

1-1/2-2 control scheme

1-2/2-1 control scheme
Control-loop Interactions

- Process interactions may induce undesirable interactions between two or more control loops.

Example: 2 x 2 system

Change in $U_1$ has two effects on $Y_1$

1) direct effect: $U_1 \rightarrow G_{p11} \rightarrow Y_1$

2) indirect effect:

$$U_1 \rightarrow G_{p21} \rightarrow Y_2 \rightarrow G_{c2} \rightarrow U_2 \rightarrow G_{p12} \rightarrow Y_1$$
• Control loop interactions are due to the presence of a third feedback loop.

Example: 1-1/2-2 pairing

The hidden feedback control loop (in dark lines)

• Problems arising from control loop interactions
  i. Closed-loop system may become destabilized.
  ii. Controller tuning becomes more difficult.
Block Diagram Analysis

For the multiloop control configuration, the transfer function between a controlled and a manipulated variable depends on whether the other feedback control loops are open or closed.

**Example: 2 x 2 system, 1-1/2 -2 pairing**

From block diagram algebra we can show

\[
\frac{Y_1(s)}{U_1(s)} = G_{p11}(s) \quad \text{(second loop open)}
\]

\[
\frac{Y_1(s)}{U_1(s)} = G_{p11} - \frac{G_{p12}G_{p21}G_{c2}}{1 + G_{c2}G_{p22}} \quad \text{(second loop closed)}
\]

Note that the last expression contains \( G_{c2} \).

\[\rightarrow\] The two controllers should not be tuned independently.
Example: Empirical model of a distillation column

\[
\begin{bmatrix}
X_D(s) \\
X_B(s)
\end{bmatrix} = \begin{bmatrix}
\frac{12.8e^{-s}}{16.7s + 1} & -\frac{18.9e^{-3s}}{21s + 1} \\
\frac{6.6e^{-7s}}{10.9s + 1} & -\frac{19.4e^{-3s}}{14.4s + 1}
\end{bmatrix}
\begin{bmatrix}
R(s) \\
S(s)
\end{bmatrix}
\]

Single-loop ITAE tuning

<table>
<thead>
<tr>
<th>Pairing</th>
<th>$K_c$</th>
<th>$\tau_I$</th>
</tr>
</thead>
<tbody>
<tr>
<td>$x_D$ - $R$</td>
<td>0.604</td>
<td>16.37</td>
</tr>
<tr>
<td>$x_B$ - $S$</td>
<td>-0.127</td>
<td>14.46</td>
</tr>
</tbody>
</table>

$x_D$ set-point response

$x_B$ set-point response
Closed-Loop Stability

- Relation between controlled variables and set-points
  
  \[ Y_1 = \Gamma_{11} Y_{sp1} + \Gamma_{12} Y_{sp2} \]
  
  \[ Y_2 = \Gamma_{21} Y_{sp1} + \Gamma_{22} Y_{sp2} \]

- Closed-loop transfer functions
  
  \[ \Gamma_{11} = \frac{G_{c1}G_{p11} + G_{c1}G_{c2}(G_{p11}G_{p22} - G_{p12}G_{p21})}{\Delta(s)} \]
  
  \[ \Gamma_{12} = \frac{G_{c2}G_{p12}}{\Delta(s)} \]
  
  \[ \Gamma_{21} = \frac{G_{c1}G_{p21}}{\Delta(s)} \]
  
  \[ \Gamma_{22} = \frac{G_{c2}G_{p22} + G_{c1}G_{c2}(G_{p11}G_{p22} - G_{p12}G_{p21})}{\Delta(s)} \]

  where  \[ \Delta(s) = (1 + G_{c1}G_{p11})(1 + G_{c2}G_{p22}) - G_{c1}G_{c2}G_{p12}G_{p21} \]

- Characteristic equation
  
  \[ (1 + G_{c1}G_{p11})(1 + G_{c2}G_{p22}) - G_{c1}G_{c2}G_{p12}G_{p21} = 0 \]
**Example:** Two P controllers are used to control the process

\[
G_p(s) = \begin{bmatrix}
\frac{2}{10s + 1} & \frac{1.5}{s + 1} \\
\frac{1.5}{s + 1} & \frac{2}{10s + 1}
\end{bmatrix}
\]

Stable region for \(K_{c1}\) and \(K_{c2}\)

1-1/2-2 pairing

1-2/2-1 pairing
Pairing of Controlled and Manipulated Variables

- Control of distillation column
  - Controlled variables: $x_D, x_B, P, h_D, h_B$
  - Manipulated variables: $D, B, R, Q_D, Q_B$

Possible multiloop control strategies

$= 5! = 120$
• One of the practical pairing

\[ R \rightarrow x_D \]
\[ Q_B \rightarrow x_B \]
\[ Q_D \rightarrow P \]
\[ D \rightarrow h_D \]
\[ B \rightarrow h_B \]
Relative Gain Array (RGA)
(Bristol, 1966)

- Provides two types of useful information:
  1. Measure of process interactions
  2. Recommendation about best pairing of controlled and manipulated variables.

- Requires knowledge of steady-state gains but not process dynamics.
Example of RGA Analysis: 2 x 2 system

- Steady-state process model

\[ y_1 = K_{11}u_1 + K_{12}u_2 \]
\[ y_2 = K_{21}u_1 + K_{22}u_2 \]

or \[ y = Ku \]

The RGA, \( \Lambda \), is defined as:

\[
\Lambda = \begin{bmatrix}
\lambda_{11} & \lambda_{12} \\
\lambda_{21} & \lambda_{22}
\end{bmatrix}
\]

where the relative gain, \( \lambda_{ij} \), relates the \( i^{th} \) controlled variable and the \( j^{th} \) manipulated variable

\[
\lambda_{ij} \triangleq \left( \frac{\partial y_i}{\partial u_j} \right)_u = \frac{\text{open-loop gain}}{\text{closed-loop gain}}
\]

\[
\left( \frac{\partial y_i}{\partial u_j} \right)_u : \text{partial derivative evaluated with all of the manipulated variables except } u_j \text{ held constant (} K_{ij} \text{)}
\]

\[
\left( \frac{\partial y_i}{\partial u_j} \right)_y : \text{partial derivative evaluated with all of the controlled variables except } y_i \text{ held constant}
\]
Scaling Properties:

i. $\lambda_{ij}$ is dimensionless

ii. $\sum_i \lambda_{ij} = \sum_j \lambda_{ij} = 1$

For a 2 x 2 system,

$$\lambda_{11} = \frac{1}{1 - \frac{K_{12}K_{21}}{K_{11}K_{22}}}, \quad \lambda_{12} = 1 - \lambda_{11} = \lambda_{21}$$

$$\Lambda = \begin{bmatrix} \lambda & 1 - \lambda \\ 1 - \lambda & \lambda \end{bmatrix} \quad (\lambda = \lambda_{11})$$

Recommended Controller Pairing

It corresponds to the $\lambda_{ij}$ which have the largest positive values that are closest to one.
**In general:**

1. Pairings which correspond to negative pairings should not be selected.
2. Otherwise, choose the pairing which has $\lambda_{ij}$ closest to one.

**Examples:**

<table>
<thead>
<tr>
<th>Process Gain Matrix, $K$ :</th>
<th>Relative Gain Array, $\Lambda :$</th>
</tr>
</thead>
<tbody>
<tr>
<td>$\begin{bmatrix} K_{11} &amp; 0 \ 0 &amp; K_{22} \end{bmatrix}$</td>
<td>$\begin{bmatrix} 1 &amp; 0 \ 0 &amp; 1 \end{bmatrix}$</td>
</tr>
<tr>
<td>$\begin{bmatrix} 0 &amp; K_{12} \ K_{21} &amp; 0 \end{bmatrix}$</td>
<td>$\begin{bmatrix} 0 &amp; 1 \ 1 &amp; 0 \end{bmatrix}$</td>
</tr>
<tr>
<td>$\begin{bmatrix} K_{11} &amp; K_{12} \ 0 &amp; K_{22} \end{bmatrix}$</td>
<td>$\begin{bmatrix} 1 &amp; 0 \ 0 &amp; 1 \end{bmatrix}$</td>
</tr>
<tr>
<td>$\begin{bmatrix} K_{11} &amp; 0 \ K_{21} &amp; K_{22} \end{bmatrix}$</td>
<td>$\begin{bmatrix} 1 &amp; 0 \ 0 &amp; 1 \end{bmatrix}$</td>
</tr>
</tbody>
</table>
For 2 x 2 systems:

\[
y_1 = K_{11}u_1 + K_{12}u_2 \\
y_2 = K_{21}u_1 + K_{22}u_2
\]

\[
\lambda_{11} = \frac{1}{1 - \frac{K_{12}K_{21}}{K_{11}K_{22}}}, \quad \lambda_{12} = 1 - \lambda_{11} = \lambda_{21}
\]

**Example 1:**

\[
K = \begin{bmatrix} K_{11} & K_{12} \\ K_{21} & K_{22} \end{bmatrix} = \begin{bmatrix} 2 & 1.5 \\ 1.5 & 2 \end{bmatrix}
\]

\[
\Lambda = \begin{bmatrix} 2.29 & -1.29 \\ -1.29 & 2.29 \end{bmatrix}
\]

∴ Recommended pairing is \(Y_1\) and \(U_1\), \(Y_2\) and \(U_2\).

**Example 2:**

\[
K = \begin{bmatrix} -2 & 1.5 \\ 1.5 & 2 \end{bmatrix} \Rightarrow \Lambda = \begin{bmatrix} 0.64 & 0.36 \\ 0.36 & 0.64 \end{bmatrix}
\]

∴ Recommended pairing is \(Y_1\) with \(U_1\) and \(Y_2\) with \(U_2\).
**EXAMPLE: Blending System**

Controlled variables: \( w \) and \( x \)

Manipulated variables: \( w_A \) and \( w_B \)

Steady-state model:

\[
\begin{align*}
  w &= w_A + w_B \\
  xw &= w_A \\
  \Rightarrow x &= \frac{w_A}{w_A + w_B}
\end{align*}
\]

The RGA is:

\[
A = \begin{bmatrix}
  w & w_b \\
  x & 1-x \\
  1-x & x
\end{bmatrix}
\]

Note that each relative gain is between 0 and 1. The recommended controller pairing depends on the desired product composition \( x \).

For \( x = 0.4 \), \( w-w_B / x-w_A \) (large interactions)

For \( x = 0.9 \), \( w-w_A / x-w_B \) (small interactions)
RGA for Higher-Order Systems

For a \( n \times n \) system,
\[
\Lambda = \begin{bmatrix}
\lambda_{11} & \lambda_{12} & \cdots & \lambda_{1n} \\
\lambda_{21} & \lambda_{22} & \cdots & \lambda_{2n} \\
\vdots & \vdots & \ddots & \vdots \\
\lambda_{n1} & \lambda_{n1} & \cdots & \lambda_{nn}
\end{bmatrix}
\]
(18–25)

Each \( \lambda_{ij} \) can be calculated from the relation,
\[
\lambda_{ij} = K_{ij} H_{ij}
\]
(18–37)

where \( K_{ij} \) is the \((i,j)\)-element of the steady-state gain \( K \) matrix,
\( H_{ij} \) is the \((i,j)\)-element of the \( H = \left(K^{-1}\right)^T \).

In matrix form,
\[
\Lambda = K \otimes H
\]
\( \otimes \): Schur product
(element by element multiplication)

Note: \( \Lambda \neq KH \)
Example: Hydrocracker

The RGA for a hydrocracker has been reported as,

\[
\Lambda = \begin{bmatrix}
0.931 & 0.150 & 0.080 & -0.164 \\
-0.011 & -0.429 & 0.286 & 1.154 \\
-0.135 & 3.314 & -0.270 & -1.910 \\
0.215 & -2.030 & 0.900 & 1.919 \\
\end{bmatrix}
\]

Recommended controller pairing?
An important disadvantage of RGA approach is that it ignores process dynamics

**Example:**

\[
G_p(s) = \begin{bmatrix}
\frac{-2e^{-s}}{10s+1} & \frac{1.5e^{-s}}{s+1} \\
\frac{1.5e^{-s}}{s+1} & \frac{2}{10s+1}
\end{bmatrix}
\]

\[\lambda_{11} = 0.64\]

Recommended controller pairing?
Singular Value Analysis

- **Any real** $m \times n$ **matrix can be factored as**, 
  $$K = W \Sigma V^T$$

- **Matrix** $\Sigma$ **is a diagonal matrix of singular values**:
  $$\Sigma = \text{diag}(\sigma_1, \sigma_2, \ldots, \sigma_r)$$

- The singular values are the positive square roots of the eigenvalues of $K^T K$ ($r$ = the rank of $K^T K$).

- The columns of matrices $W$ and $V$ are **orthonormal**. Thus, 
  $$WW^T = I \quad \text{and} \quad VV^T = I$$

- Can calculate $\Sigma$, $W$, and $V$ using MATLAB command, **svd**.

- **Condition number** ($CN$) is defined to be the ratio of the largest to the smallest singular value,
  $$CN \triangleq \frac{\sigma_1}{\sigma_r}$$

- A large value of $CN$ indicates that $K$ is ill-conditioned.
Condition Number

- CN is a measure of sensitivity of the matrix properties to changes in individual elements.
- Consider the RGA for a 2x2 process,

\[ K = \begin{bmatrix} 1 & 0 \\ 10 & 1 \end{bmatrix} \Rightarrow \Lambda = I \]

- If \( K_{12} \) changes from 0 to 0.1, then \( K \) becomes a singular matrix, which corresponds to a process that is difficult to control.
- RGA and SVA used together can indicate whether a process is easy (or difficult) to control.

\[ \Sigma (K) = \begin{bmatrix} 10.1 & 0 \\ 0 & 0.1 \end{bmatrix} \quad \text{CN} = 101 \]

- \( K \) is poorly conditioned when CN is a large number (e.g., > 10). Thus small changes in the model for this process can make it very difficult to control.
Selection of Inputs and Outputs

- Arrange the singular values in order of largest to smallest and look for any $\sigma_i/\sigma_{i-1} > 10$; then one or more inputs (or outputs) can be deleted.
- Delete one row and one column of $K$ at a time and evaluate the properties of the reduced gain matrix.

- Example:

$$K = \begin{bmatrix}
0.48 & 0.90 & -0.006 \\
0.52 & 0.95 & 0.008 \\
0.90 & -0.95 & 0.020
\end{bmatrix}$$
The RGA is:

\[ W = \begin{bmatrix}
0.5714 & 0.3766 & 0.7292 \\
0.6035 & 0.4093 & -0.6843 \\
-0.5561 & 0.8311 & 0.0066
\end{bmatrix} \]

\[ \Sigma = \begin{bmatrix}
1.618 & 0 & 0 \\
0 & 1.143 & 0 \\
0 & 0 & 0.0097
\end{bmatrix} \]

\[ V = \begin{bmatrix}
0.0541 & 0.9984 & 0.0151 \\
0.9985 & -0.0540 & -0.0068 \\
-0.0060 & 0.0154 & -0.9999
\end{bmatrix} \]

CN = 166.5 (\( \sigma_1/\sigma_3 \))

The RGA is:

\[ \Lambda = \begin{bmatrix}
-2.4376 & 3.0241 & 0.4135 \\
1.2211 & -0.7617 & 0.5407 \\
2.2165 & -1.2623 & 0.0458
\end{bmatrix} \]

Preliminary pairing: \( y_1-u_2, \ y_2-u_3, \ y_3-u_1. \)

CN suggests only two output variables can be controlled. Eliminate one input and one output (3x3→2x2).
<table>
<thead>
<tr>
<th>Pairing Number</th>
<th>Controlled Variables</th>
<th>Manipulated Variables</th>
<th>CN</th>
<th>λ</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>( y_1, y_2 )</td>
<td>( u_1, u_2 )</td>
<td>184</td>
<td>39.0</td>
</tr>
<tr>
<td>2</td>
<td>( y_1, y_2 )</td>
<td>( u_1, u_3 )</td>
<td>72.0</td>
<td>0.552</td>
</tr>
<tr>
<td>3</td>
<td>( y_1, y_2 )</td>
<td>( u_2, u_3 )</td>
<td>133</td>
<td>0.558</td>
</tr>
<tr>
<td>4</td>
<td>( y_1, y_3 )</td>
<td>( u_2, u_1 )</td>
<td><strong>1.51</strong></td>
<td>0.640</td>
</tr>
<tr>
<td>5</td>
<td>( y_1, y_3 )</td>
<td>( u_1, u_3 )</td>
<td>69.4</td>
<td>0.640</td>
</tr>
<tr>
<td>6</td>
<td>( y_1, y_3 )</td>
<td>( u_2, u_3 )</td>
<td>139</td>
<td>1.463</td>
</tr>
<tr>
<td>7</td>
<td>( y_2, y_3 )</td>
<td>( u_2, u_1 )</td>
<td><strong>1.45</strong></td>
<td>0.634</td>
</tr>
<tr>
<td>8</td>
<td>( y_2, y_3 )</td>
<td>( u_1, u_3 )</td>
<td>338</td>
<td>3.25</td>
</tr>
<tr>
<td>9</td>
<td>( y_2, y_3 )</td>
<td>( u_2, u_3 )</td>
<td>67.9</td>
<td>0.714</td>
</tr>
</tbody>
</table>

**Table 18.3** CN and \( \lambda \) for Different \( 2 \times 2 \) Pairings, Example 18.7
Matrix Notation for Multiloop Control Systems

**Single loop**

CLTF \[ Y = \frac{G_p G_c}{1+G_p G_c} Y_{sp} \]

**Multi-loop**

\[ Y = \left( I + G_p G_c \right)^{-1} G_p G_c Y_{sp} \]

- \( Y \): (n x 1) vector of control variables
- \( Y_{sp} \): (n x 1) vector of set-points
- \( G_p \): (n x n) matrix of process transfer functions
- \( G_c \): (n x n) diagonal matrix of controller transfer functions

**Characteristic equation**

- Single loop: \( 1+G_p G_c = 0 \)
- Multi-loop: \( \det \left( I + G_p G_c \right) = 0 \)
Tuning of Multiloop PID Control Systems

- **Detuning method**
  - Each controller is first designed, ignoring process interactions
  - Then interactions are taken into account by detuning each controller
    - More conservative controller settings (decrease controller gain, increase integral time)
  - **Tyreus-Luyben (TL) tuning**

<table>
<thead>
<tr>
<th>Ziegler-Nichols</th>
<th>$K_c$</th>
<th>$\tau_I$</th>
<th>$\tau_D$</th>
</tr>
</thead>
<tbody>
<tr>
<td>P</td>
<td>$0.5K_{cu}$</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>PI</td>
<td>$0.45K_{cu}$</td>
<td>$P_u/1.2$</td>
<td>—</td>
</tr>
<tr>
<td>PID</td>
<td>$0.6K_{cu}$</td>
<td>$P_u/2$</td>
<td>$P_u/8$</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Tyreus-Luyben†</th>
<th>$K_c$</th>
<th>$\tau_I$</th>
<th>$\tau_D$</th>
</tr>
</thead>
<tbody>
<tr>
<td>PI</td>
<td>$0.31K_{cu}$</td>
<td>$2.2P_u$</td>
<td>—</td>
</tr>
<tr>
<td>PID</td>
<td>$0.45K_{cu}$</td>
<td>$2.2P_u$</td>
<td>$P_u/6.3$</td>
</tr>
</tbody>
</table>

† Luyben and Luyben (1997).
Biggest log-modulus tuning (BLT) method (Luyben, 1986)

- **Log-modulus**: a robustness measure of control systems
  - **Single loop**
    \[
    L_c = 20 \log \left| \frac{G_p G_c}{1 + G_p G_c} \right| = 20 \log \left| \frac{G}{1 + G} \right|
    \]
    \[
    L_c^{\text{max}} = \max_{\omega} L_c = \max_{\omega} \left\{ 20 \log \left| \frac{G}{1 + G} \right| \right\}
    \]
    A specification of \( L_c^{\text{max}} = 2 \text{ dB} \) has been suggested.
  - **Multi-loop**
    Define \( W = -1 + \det(I + G_p G_c) \)
    \[
    L_c = 20 \log \left| \frac{W}{1 + W} \right|
    \]
    Luyben suggest that \( L_c^{\text{max}} = \max_{\omega} L_c = 2n \)
    where \( n \) is the dimension of the multivariable system.
Tuning Procedure of BLT method

1. Calculate Z-N PI controller settings for each control loop
   \[ K_{c,ZN} = 0.45 K_{cu}, \quad \tau_{i,ZN} = P_u / 1.2 \]

2. Assume a factor \( F \); typical values between 2 and 5

3. Calculate new values of controller parameters by
   \[ K_{ci} = \frac{K_{ci,ZN}}{F}, \quad \tau_{i} = F \tau_{i,ZN} ; \quad i = 1, 2, \ldots, n \]  \hspace{1cm} (detuning)

4. Compute \( W = -1 + \det \left( I + G_p G_c \right) \) for \( 0 \leq \omega < \infty \)
   for example, 2x2 system
   \[ \det \left( I + G_p G_c \right) = 1 + G_{c1} G_{p11} + G_{c2} G_{p22} + G_{c1} G_{c2} \left( G_{p11} G_{p22} - G_{p12} G_{p21} \right) \]

5. Determine
   \[ L_c^{\text{max}} = \max_{\omega} \left\{ 20 \log \left| \frac{W}{1+W} \right| \right\} \]

6. If \( L_c^{\text{max}} \neq 2n \), select a new value of \( F \) and return to step 2 until \( L_c^{\text{max}} = 2n \)
Multiloop IMC Controller

- Design IMC controller based in diagonal process transfer functions

\[
G_p = \begin{bmatrix}
G_{p11} & \cdots & G_{p1n} \\
\vdots & \ddots & \vdots \\
G_{pn1} & \cdots & G_{pnn}
\end{bmatrix}
\]

- The IMC controller is designed as

\[
G_c = \text{diag}[G_{c1} \quad G_{c2} \quad \cdots \quad G_{cn}]
\]

with

\[
G_{ci} = G_{pii}^{-1} \cdot f_i \quad i = 1, 2, \cdots, n
\]

- Since the off-diagonal terms of \( G_p \) have been dropped, modeling error are always present.
Alternative Strategies for Dealing with Undesirable Control Loop Interactions

1. "Detune" one or more FB controllers.
2. Select different manipulated or controlled variables. e.g., nonlinear functions of original variables
3. Use a decoupling control scheme.
4. Use some other type of multivariable control scheme.

Decoupling Control Systems

- **Basic Idea**: Use additional controllers (decoupler) to compensate for process interactions and thus reduce control loop interactions

- Ideally, decoupling control allows setpoint changes to affect only the desired controlled variables.

- Typically, decoupling controllers are designed using a simple process model (e.g., a steady-state model or transfer function model)
A Decoupling Control System

decoupler
Decoupler Design Equations

We want cross-controller, \( T_{12} \), to cancel the effect of \( U_2 \) on \( Y_1 \). Thus, we would like:

\[
G_{p11}U_{12} + G_{p12}U_{22} = 0
\]

or

\[
G_{p11}T_{12}U_{22} + G_{p12}U_{22} = 0
\]

Because \( U_{22} \neq 0 \) in general, then

\[
T_{12} = -\frac{G_{p12}}{G_{p11}}
\]

Similarly, we want \( T_{12} \) to cancel the effect of \( U_1 \) on \( Y_2 \). Thus, we require that,

\[
G_{p22}T_{21}U_{11} + G_{p21}U_{11} = 0
\]

\[
T_{21} = -\frac{G_{p21}}{G_{p22}}
\]

Compare with the design equations for feedforward control based on block diagram analysis.
Variations on a Theme

1. **Partial Decoupling:**
   Use only one “cross-controller.”

2. **Static Decoupling:**
   Design to eliminate Steady-State interactions
   Ideal decouplers are merely gains:

   \[
   T_{12} = -\frac{K_{p12}}{K_{p11}}
   \]

   \[
   T_{21} = -\frac{K_{p21}}{K_{p22}}
   \]

3. **Nonlinear Decoupling**
   Appropriate for nonlinear processes.
Wood-Berry Distillation Column Model
(methanol-water separation)
Wood-Berry Distillation Column Model

\[ \begin{bmatrix} y_1(s) \\ y_2(s) \end{bmatrix} = \begin{bmatrix} \frac{12.8e^{-s}}{16.7s+1} & -\frac{18.9e^{-3s}}{21s+1} \\ \frac{6.6e^{-7s}}{10.9s+1} & -\frac{19.4e^{-3s}}{14.4s+1} \end{bmatrix} \begin{bmatrix} u_1(s) \\ u_2(s) \end{bmatrix} \]

(18–12)

where:

\[ y_1 = x_D = \text{distillate composition, } \%\text{MeOH} \]
\[ y_2 = x_B = \text{bottoms composition, } \%\text{MeOH} \]
\[ u_1 = R = \text{reflux flow rate, lb/min} \]
\[ u_1 = S = \text{reflux flow rate, lb/min} \]
Figure 19.13. An experimental application of decoupling (noninteracting) control to a distillation column [3].